Iterative detection for V-BLAST MIMO communication systems based on expectation maximisation algorithm

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By applying the expectation maximisation algorithm to the maximum likelihood detection of layered space-time codes, the conditional loglikelihood of a single layer is iteratively maximised, rather than maximising the intractable likelihood function of all layers. Computer simulations demonstrate the efficiency of the proposed detection scheme.

System description: Consider the system of N transmit and M receive antennas. The single data stream in the input is demultiplexed into Nsubstreams, and each substream is modulated independently then transmitted over a rich-scattering wireless channel to M antennas. Each antenna receives signals transmitted from the entire N transmit antennas. The transmission is performed by burst of length l. We adopt a quasi-static approximation of the fading channel, i.e. the channel remains unchanged during each burst. For simplicity of presentation, we consider a BPSK modulation for each layer. The received signal at each instant time can be written as

$$\mathbf{y} = \mathbf{H}\mathbf{b} + \mathbf{z} \tag{1}$$

where $\mathbf{z} = [z_1 z_2 \cdots z_M]^T$ is assumed to be an symmetric i.i.d. complex Gaussian noise vector with zero mean and unit variance, $\mathbf{b} = [b_1 b_2 \cdots b_N]^T$ is the transmitted signal where $\mathbf{b} \in 2\{-1, 1\}^N$, $\mathbf{y} = [y_1 y_2 \cdots y_M]^T$ is the received signal, and $H_{M \times N}$ is the channel matrix. In rich-scattering environments, the elements of the channel matrix can be modelled as symmetric i.i.d. complex Gaussian random variables with zero mean and variance equal to γ/N . From the above definitions, it is clear that γ represents the average SNR at each receive antenna.

The maximisation of the conditional probability density of the received vector in (1) is equivalent to the following nonlinear optimisation problem:

$$\hat{\mathbf{b}} = \arg \max_{\mathbf{b} \in \{-1,1\}^N} [2y^H \mathbf{H} \mathbf{b} - \mathbf{b}^H \mathbf{H}^H \mathbf{H} \mathbf{b}]$$
(2)

Since the optimisation in (2) is performed over $\{-1, 1\}^N$, its computational complexity is exponentially in *N* and thus becomes prohibitive for even a moderate number of layers. By using the EM algorithm, we decompose the *N*-dimensional maximisation problem in (2) into *N* one-dimensional maximisation problems. First, we define the complete data set \mathbf{x}_k , as required by the EM algorithm, as follows:

$$\mathbf{x}_{\mathbf{k}} = \mathbf{h}_{\mathbf{k}} b_k + \mathbf{z}_{\mathbf{k}} \tag{3}$$

where $\mathbf{x}_{\mathbf{k}}$ is an *M* element column vector, b_k is the symbol bit of the *k*th layer, $\mathbf{z}_{\mathbf{k}}$ is the complex *M*-dimensional AWGN, the components of which are statistically independent with identical power σ_k^2 , and $\mathbf{h}_{\mathbf{k}}$ is the *k*th column of **H**. From (1) and (3), we have

$$\sum_{k=1}^{N} \sigma_k^2 = \sigma^2 \tag{4}$$

and

$$\mathbf{y} = \sum_{k=1}^{N} \mathbf{x}_{\mathbf{k}}$$
(5)

where $\sigma^2 = 1$. In [1] it is proven that the maximisation of the loglikelihood function

$$L(\mathbf{b}) = \frac{-1}{2} (\mathbf{y} - \mathbf{H}\mathbf{b})^{H} (\mathbf{y} - \mathbf{H}\mathbf{b})$$
(6)

is equivalent to the maximisation of the function $U(\mathbf{b}, \mathbf{\dot{b}})$ over \mathbf{b} , where $U(\mathbf{b}, \mathbf{\dot{b}})$ is:

$$U(\mathbf{b}, \mathbf{\dot{b}}) = E[\log(f_{\mathbf{x}}(\mathbf{x}; \mathbf{b}))|\mathbf{y}, \mathbf{b} = \mathbf{\dot{b}}]$$
(7)

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and

$$\boldsymbol{f}_{\mathbf{x}}(\mathbf{x}; \mathbf{b}) = \frac{1}{(2\pi)^{N/2} \prod_{k=1}^{N} \sigma_{k}} \times \exp\left(\sum_{k=1}^{N} \frac{-1}{2\sigma_{k}^{2}} (\mathbf{x}_{\mathbf{k}} - \mathbf{h}_{\mathbf{k}} b_{k})^{H} (\mathbf{x}_{\mathbf{k}} - \mathbf{h}_{\mathbf{k}} b_{k})\right)$$
(8)

In (7–8), $\mathbf{x} = [\mathbf{x}_1^T \mathbf{x}_2^T \mathbf{x}_3^T \cdots \mathbf{x}_N^T]^T$ and $\mathbf{\dot{b}}$ is an estimate of \mathbf{b} . The iterative structure of the detection algorithm is straightforward. To decrease the probability of error, in *n*th iteration, $U(\mathbf{b}, \mathbf{b}^{(n)})$ is maximised over \mathbf{b} to obtain a new estimate of $\mathbf{b}^{(n+1)}$. After each iteration, the likelihood function increases [1]. Now an analytical solution to the maximisation of $U(\mathbf{b}, \mathbf{b}^{(n)})$ is derived to reduce the *N*-dimensional maximisation problem into *N* one-dimensional maximisation problems. We start by expanding the log-likelihood function

$$\log_e f_{\mathbf{x}}(\mathbf{x}; \mathbf{b}) = \sum_{k=1}^{N} \frac{-1}{2\sigma_k^2} (\mathbf{x}_k^H \mathbf{x}_k + b_k^2 \mathbf{h}_k^H \mathbf{h}_k - b_k \mathbf{h}_k^H \mathbf{x}_k - b_k \mathbf{x}_k^H \mathbf{h}_k) + C$$
(9)

where C is a constant. We define $\mathbf{A} = [b_1 \mathbf{h}_1^T / \sigma_1^2 \ b_2 \mathbf{h}_2^T / \sigma_2^2 \dots b_N \ \mathbf{h}_N^T / \sigma_N^2]^T$ and since $b_k^2 = 1$, (9) can be simplified to

$$\log_e f_{\mathbf{x}}(\mathbf{x}; \mathbf{b}) = g(\mathbf{x}) + 2Re(\mathbf{A}^H \mathbf{x})$$
(10)

Ignoring the first term which has no effect on the maximisation process, and substituting (10) into (7), we will have

$$U(\mathbf{b}, \mathbf{b}^{(n)}) = Re(\mathbf{A}^{H}E[\mathbf{x}|\mathbf{y}, \mathbf{b}^{(n)}])$$
(11)

It can be easily shown that

$$E[\mathbf{x}|\mathbf{y}, \mathbf{b}^{(n)}] = \frac{1}{\sigma^2} [\sigma_1^2 (\mathbf{y} - \mathbf{H} \mathbf{b}^{(n)})^T \sigma_2^2 (\mathbf{y} - \mathbf{H} \mathbf{b}^{(n)})^T \dots \sigma_N^2 (\mathbf{y} - \mathbf{H} \mathbf{b}^{(n)})^T]^T + [b_1^{(n)} \mathbf{h}_1^T b_2^{(n)} \mathbf{h}_2^T \dots b_N^{(n)} \mathbf{h}_N^T]^T$$
(12)

Hence,

$$U(\mathbf{b}, \mathbf{b}^{(n)}) = \sum_{k=1}^{N} Re\left(b_k \frac{\mathbf{h}_k^H}{\sigma_k^2} \left[b_k^{(n)} \mathbf{h}_k + \frac{\sigma_k^2}{\sigma^2} (\mathbf{y} - \mathbf{H} \mathbf{b}^{(n)})\right]\right)$$
(13)

Obviously, to maximise the whole sum, it is sufficient to maximise each term in the sum, separately. Since b_k 's can only take the values +1 and -1, the following decisions on b_k 's maximise (13):

$$b_{k}^{(n+1)} = sign\left\{ Re\left(b_{k}^{(n)} \mathbf{h}_{\mathbf{k}}^{H} \mathbf{h}_{\mathbf{k}} + \frac{\sigma_{k}^{2}}{\sigma^{2}} \mathbf{h}_{\mathbf{k}}^{H} (\mathbf{y} - \mathbf{H} \mathbf{b}^{(n)}) \right) \right\}$$
(14)



Fig. 1 Plot of average BER against SNR at each receive antenna

Simulation results: We have evaluated the performance of the proposed algorithm by simulation. We considered N=8, M=8. Also, we assumed a block fading model where the channel matrix remains constant within each burst, but independently changes from burst to burst. Fig. 1 compares the performance of our algorithm with that of the nulling and cancelling with optimal ordering. As can be realised, our proposed EM-based algorithm substantially outperforms

the original nulling and cancelling with optimal ordering scheme proposed in [2], over the whole SNR range.

© IEE 2004 *I February 2004 Electronics Letters* online no: 20040454 doi: 10.1049/el:20040454

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